

# ECE171A: Linear Control System Theory

## Lecture 4: ODE Solutions

Nikolay Atanasov  
natanasov@ucsd.edu

**UC San Diego**  
**JACOBS SCHOOL OF ENGINEERING**  
Electrical and Computer Engineering

# Outline

Examples

Linear Properties of LTI Systems

LTI ODE Solution

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## Examples

Linear Properties of LTI Systems

LTI ODE Solution

# Existence and Uniqueness of ODE Solutions

- ▶ Consider the nonlinear initial value problem:

$$\dot{\mathbf{x}} = \mathbf{F}(\mathbf{x}), \quad \mathbf{x}(t_0) = \mathbf{x}_0$$

- ▶ A function  $\mathbf{s}(t)$  is a **solution** to the initial value problem on interval  $[t_0, t_f]$  if:

$$\mathbf{s}(t_0) = \mathbf{x}_0 \quad \text{and} \quad \frac{d}{dt}\mathbf{s}(t) = \mathbf{F}(\mathbf{s}(t)), \quad \forall t_0 < t < t_f$$

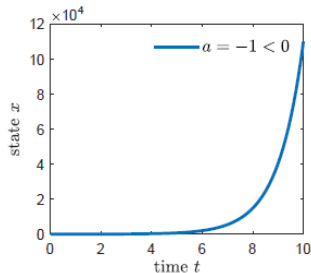
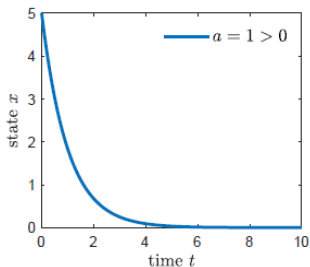
- ▶ If the function  $\mathbf{F}(\mathbf{x})$  is well-behaved (Lipschitz continuous), then the initial value problem has a **unique solution**
- ▶ In general, a nonlinear initial value problem:
  - ▶ may not have a unique solution (see Example 5.3:  $\dot{x} = 2\sqrt{x}$ )
  - ▶ may not have a solution (see Example 5.2:  $\dot{x} = x^2$ )

## Example 1: Scalar System

- ▶ Consider the scalar system:

$$\dot{x} = -ax, \quad x(0) = x_0$$

- ▶ Its unique solution is  $x(t) = e^{-at}x_0$



## Example 2: Decoupled Two-dimensional System

- ▶ Consider the system:

$$\frac{d}{dt} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \underbrace{\begin{bmatrix} -a & 0 \\ 0 & -b \end{bmatrix}}_A \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

- ▶ Its unique solution is:

$$x_1(t) = e^{-at} x_1(0), \quad x_2(t) = e^{-bt} x_2(0)$$

- ▶ Note the vector form of the solution:

$$\mathbf{x}(t) = \begin{bmatrix} e^{-at} & 0 \\ 0 & e^{-bt} \end{bmatrix} \mathbf{x}(0)$$

## Example 3: Double Integrator

- ▶ Consider the system with constant  $a \in \mathbb{R}$ :

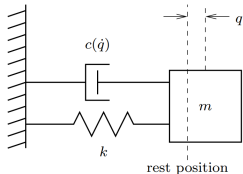
$$\frac{d}{dt} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \underbrace{\begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix}}_A \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ a \end{bmatrix}$$

- ▶ Interpret the system state as position  $x_1(t)$  and velocity  $x_2(t)$
- ▶ Determine the velocity solution first
- ▶ The unique solution is:

$$x_1(t) = x_1(0) + x_2(0)t + \frac{1}{2}at^2$$

$$x_2(t) = x_2(0) + at$$

## Example 4: Damped Oscillator (Spring-mass System)



$m$  = mass

$F$  = External force

$c$  = friction (damper)

$k$  = spring stiffness

$q$  = position

- **System model:** from Newton's second law:

$$m\ddot{q} + c\dot{q} + kq = F$$

- **Free response:** let  $F = 0$ :

$$m\ddot{q} + c\dot{q} + kq = 0 \quad \Rightarrow \quad \ddot{q} + \frac{c}{m}\dot{q} + \frac{k}{m}q = 0$$

- Introduce damping ratio  $\zeta$  and natural frequency  $\omega_0$  parameters:

$$2\zeta\omega_0 = \frac{c}{m}, \quad \omega_0^2 = \frac{k}{m} \quad \Rightarrow \quad \ddot{q} + 2\zeta\omega_0\dot{q} + \omega_0^2q = 0$$



## Example 4: Damped Oscillator (Spring-mass System)

- ▶ State variables:

$$x_1 = q, \quad x_2 = \frac{\dot{q}}{\omega_0}$$

- ▶ State-space model:

$$\frac{d}{dt} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \begin{bmatrix} \omega_0 x_2 \\ -\omega_0 x_1 - 2\zeta \omega_0 x_2 \end{bmatrix} = \underbrace{\begin{bmatrix} 0 & \omega_0 \\ -\omega_0 & -2\zeta \omega_0 \end{bmatrix}}_A \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

- ▶ Assume  $\zeta < 1$  (underdamped oscillator) and define the damped frequency:

$$\omega_d = \omega_0 \sqrt{1 - \zeta^2}$$

- ▶ The unique solution is:

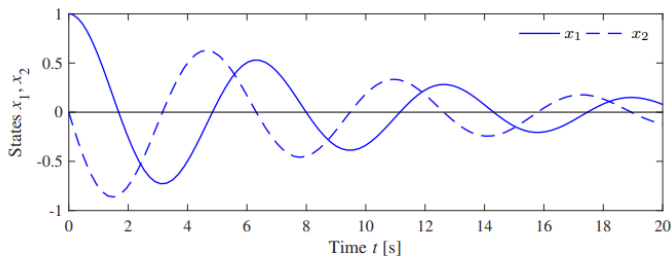
$$x_1(t) = e^{-\zeta \omega_0 t} (x_1(0) \cos(\omega_d t) + a_1 \sin(\omega_d t))$$

$$x_2(t) = e^{-\zeta \omega_0 t} (x_2(0) \cos(\omega_d t) + a_2 \sin(\omega_d t))$$

where  $a_1, a_2$  are constants depending on the initial conditions  $x_1(0), x_2(0)$ :

$$a_1 = \frac{1}{\omega_d} (\omega_0 \zeta x_1(0) + x_2(0)), \quad a_2 = -\frac{1}{\omega_d} (\omega_0^2 x_1(0) + \omega_0 \zeta x_2(0))$$

## Example 4: Damped Oscillator (Spring-mass System)



**Figure 5.1:** Response of the damped oscillator to the initial condition  $x_0 = (1, 0)$ . The solution is unique for the given initial conditions and consists of an oscillatory solution for each state, with an exponentially decaying magnitude.

# Outline

Examples

Linear Properties of LTI Systems

LTI ODE Solution

# LTI ODE System

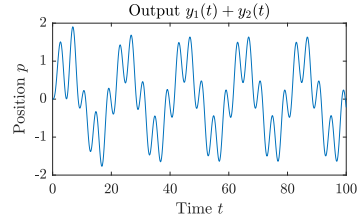
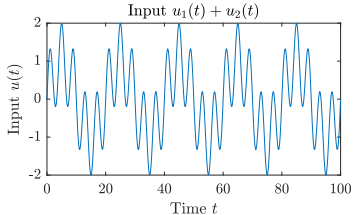
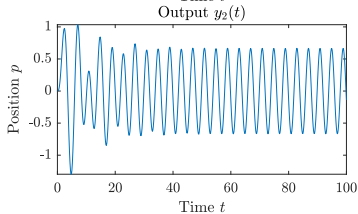
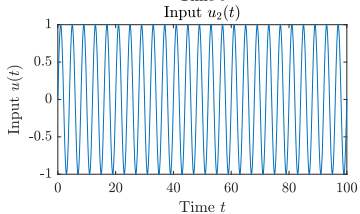
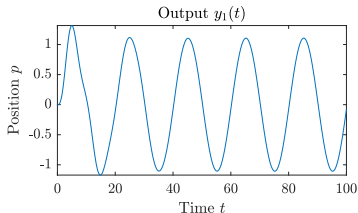
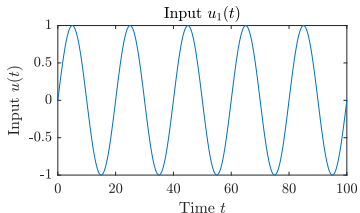
- ▶ Consider the LTI ODE system:

$$\dot{\mathbf{x}} = \mathbf{Ax} + \mathbf{Bu}$$

$$\mathbf{y} = \mathbf{Cx} + \mathbf{Du}$$

- ▶ The output  $\mathbf{y}(t)$  satisfies **linear properties**:
  - ▶ **Case 1**: Zero initial state  $\mathbf{x}(0) = \mathbf{0}$ : *the output  $\mathbf{y}(t)$  is linear in input  $\mathbf{u}(t)$*
  - ▶ **Case 2**: Zero input  $\mathbf{u}(t) \equiv 0$ : *the output  $\mathbf{y}(t)$  is linear in initial state  $\mathbf{x}(0)$*

# Case 1: Zero Initial State $x(0) = 0$



## Case 1: Zero Initial State $\mathbf{x}(0) = \mathbf{0}$

Zero initial state  $\mathbf{x}(0) = \mathbf{0}$ : the output  $\mathbf{y}(t)$  is linear in input  $\mathbf{u}(t)$

$$\begin{cases} \mathbf{u}_1(t) \rightarrow \mathbf{y}_1(t) \\ \mathbf{u}_2(t) \rightarrow \mathbf{y}_2(t) \end{cases} \implies \alpha \mathbf{u}_1(t) + \beta \mathbf{u}_2(t) \rightarrow \alpha \mathbf{y}_1(t) + \beta \mathbf{y}_2(t)$$

### Proof:

- ▶ Denote the state trajectory for  $\mathbf{u}_1(t)$  as  $\mathbf{x}_1(t)$ , and for  $\mathbf{u}_2(t)$  as  $\mathbf{x}_2(t)$ :

$$\dot{\mathbf{x}}_1(t) = \mathbf{A}\mathbf{x}_1(t) + \mathbf{B}\mathbf{u}_1(t), \quad \dot{\mathbf{x}}_2(t) = \mathbf{A}\mathbf{x}_2(t) + \mathbf{B}\mathbf{u}_2(t)$$

- ▶ Let  $\mathbf{u}(t) = \alpha \mathbf{u}_1(t) + \beta \mathbf{u}_2(t)$  and verify  $\mathbf{x}(t) = \alpha \mathbf{x}_1(t) + \beta \mathbf{x}_2(t)$  is a solution:

- ▶ **Initial condition:**  $\mathbf{x}(0) = \alpha \mathbf{x}_1(0) + \beta \mathbf{x}_2(0) = \mathbf{0}$

- ▶ **ODE:**

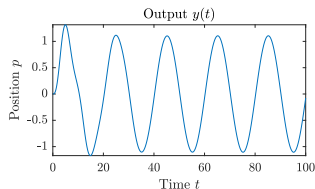
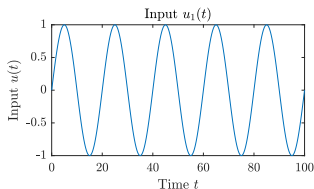
$$\begin{aligned} \dot{\mathbf{x}} &= \alpha \dot{\mathbf{x}}_1 + \beta \dot{\mathbf{x}}_2 = \alpha(\mathbf{A}\mathbf{x}_1 + \mathbf{B}\mathbf{u}_1) + \beta(\mathbf{A}\mathbf{x}_2 + \mathbf{B}\mathbf{u}_2) \\ &= \mathbf{A}(\alpha \mathbf{x}_1 + \beta \mathbf{x}_2) + \mathbf{B}(\alpha \mathbf{u}_1 + \beta \mathbf{u}_2) \\ &= \mathbf{A}\mathbf{x} + \mathbf{B}\mathbf{u} \end{aligned}$$

- ▶ Hence, the output corresponding to  $\mathbf{u} = \alpha \mathbf{u}_1 + \beta \mathbf{u}_2$  is:

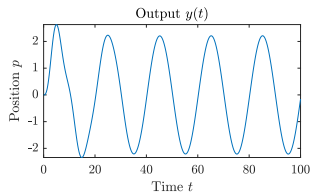
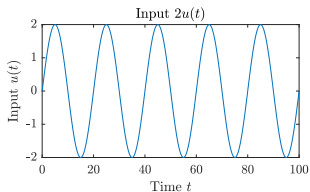
$$\begin{aligned} \mathbf{y} &= \mathbf{C}\mathbf{x} + \mathbf{D}\mathbf{u} = \mathbf{C}(\alpha \mathbf{x}_1 + \beta \mathbf{x}_2) + \mathbf{D}(\alpha \mathbf{u}_1 + \beta \mathbf{u}_2) \\ &= \alpha(\mathbf{C}\mathbf{x}_1 + \mathbf{D}\mathbf{u}_1) + \beta(\mathbf{C}\mathbf{x}_2 + \mathbf{D}\mathbf{u}_2) = \alpha \mathbf{y}_1 + \beta \mathbf{y}_2 \end{aligned}$$

## Case 1: Zero Initial State $\mathbf{x}(0) = \mathbf{0}$

- ▶ Consider an LTI ODE with zero initial condition
- ▶ Suppose that with input  $\mathbf{u}(t)$ , the output is  $\mathbf{y}(t)$



- ▶ If the input is  $2\mathbf{u}(t)$ , what is the output?



- ▶ If the input amplitude is doubled, then the output amplitude is also doubled

## Case 2: Zero Input $\mathbf{u}(t) \equiv \mathbf{0}$

Zero input  $\mathbf{u}(t) \equiv \mathbf{0}$ : the output  $\mathbf{y}(t)$  is linear in the initial state  $\mathbf{x}(0)$

$$\begin{cases} \mathbf{x}_1(0) = \boldsymbol{\xi}_1 \rightarrow \mathbf{y}_1(t) \\ \mathbf{x}_2(0) = \boldsymbol{\xi}_2 \rightarrow \mathbf{y}_2(t) \end{cases} \implies \mathbf{x}_3(0) = \alpha\boldsymbol{\xi}_1 + \beta\boldsymbol{\xi}_2 \rightarrow \alpha\mathbf{y}_1(t) + \beta\mathbf{y}_2(t)$$

**Proof:**

- ▶ Denote the state trajectory for  $\boldsymbol{\xi}_1$  as  $\mathbf{x}_1(t)$ , and for  $\boldsymbol{\xi}_2$  as  $\mathbf{x}_2(t)$ :

$$\dot{\mathbf{x}}_1(t) = \mathbf{A}\mathbf{x}_1(t), \quad \dot{\mathbf{x}}_2(t) = \mathbf{A}\mathbf{x}_2(t)$$

- ▶ Verify that  $\mathbf{x}_3(t) = \alpha\mathbf{x}_1(t) + \beta\mathbf{x}_2(t)$  is a solution:

- ▶ **Initial condition:**  $\mathbf{x}_3(0) = \alpha\mathbf{x}_1(0) + \beta\mathbf{x}_2(0) = \alpha\boldsymbol{\xi}_1 + \beta\boldsymbol{\xi}_2$

- ▶ **ODE:**

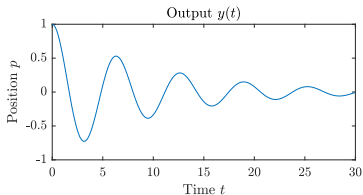
$$\dot{\mathbf{x}}_3 = \alpha\dot{\mathbf{x}}_1 + \beta\dot{\mathbf{x}}_2 = \alpha\mathbf{A}\mathbf{x}_1 + \beta\mathbf{A}\mathbf{x}_2 = \mathbf{A}(\alpha\mathbf{x}_1 + \beta\mathbf{x}_2) = \mathbf{A}\mathbf{x}_3$$

- ▶ Hence, the output corresponding to  $\mathbf{x}_3(t) = \alpha\mathbf{x}_1(t) + \beta\mathbf{x}_2(t)$  is:

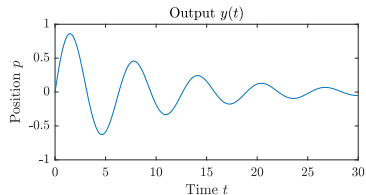
$$\mathbf{y}_3(t) = \mathbf{C}\mathbf{x}_3(t) = \mathbf{C}(\alpha\mathbf{x}_1(t) + \beta\mathbf{x}_2(t)) = \alpha\mathbf{y}_1(t) + \beta\mathbf{y}_2(t)$$



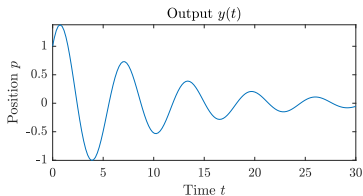
## Case 2: Zero Input $u(t) \equiv 0$



(a) Initial condition  $\mathbf{x}(0) = \xi_1$



(b) Initial condition  $\mathbf{x}(0) = \xi_2$



(c) Initial condition  $\mathbf{x}(0) = \xi_1 + \xi_2$

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## Homogeneous LTI ODE Solution

- ▶ Consider the homogeneous scalar LTI ODE:

$$\dot{x} = ax, \quad x(0) = x_0$$

- ▶ Its solution is:

$$x(t) = e^{at} x_0$$

- ▶ Consider the homogeneous vector LTI ODE:

$$\dot{\mathbf{x}} = \mathbf{A}\mathbf{x}, \quad \mathbf{x}(0) = \mathbf{x}_0$$

- ▶ What is the solution?

## Homogeneous LTI ODE Solution

### Theorem

The homogeneous vector linear time-invariant ordinary differential equation:

$$\dot{\mathbf{x}} = \mathbf{A}\mathbf{x}, \quad \mathbf{x}(t_0) = \mathbf{x}_0$$

has a unique solution:

$$\mathbf{x}(t) = e^{\mathbf{A}(t-t_0)}\mathbf{x}_0$$

### Definition

The exponential function of a matrix  $\mathbf{X} \in \mathbb{R}^{n \times n}$  is defined as:

$$e^{\mathbf{X}} = \mathbf{I} + \mathbf{X} + \frac{1}{2}\mathbf{X}^2 + \frac{1}{3!}\mathbf{X}^3 + \dots = \sum_{k=0}^{\infty} \frac{1}{k!}\mathbf{X}^k,$$

where  $\mathbf{I}$  is the  $n \times n$  identity matrix.

- ▶ **Note:** it is immediate to see that the solution  $\mathbf{x}(t) = e^{\mathbf{A}(t-t_0)}\mathbf{x}_0$  is linear in the initial condition  $\mathbf{x}_0$

## Proof

- ▶ **Initial condition:**

$$\mathbf{x}(t_0) = e^{\mathbf{A}(t_0-t_0)}\mathbf{x}_0 = e^{\mathbf{0}}\mathbf{x}_0 = \mathbf{x}_0$$

- ▶ **ODE:**

$$\begin{aligned}\frac{d}{dt}\mathbf{x}(t) &= \frac{d}{dt}\left(e^{\mathbf{A}(t-t_0)}\mathbf{x}_0\right) \\ &= \frac{d}{dt}\left(\mathbf{I} + \mathbf{A}(t-t_0) + \frac{1}{2}\mathbf{A}^2(t-t_0)^2 + \frac{1}{3!}\mathbf{A}^3(t-t_0)^3 \dots\right)\mathbf{x}_0 \\ &= \left(\mathbf{0} + \mathbf{A} + \mathbf{A}^2(t-t_0) + \frac{1}{2!}\mathbf{A}^3(t-t_0)^2 + \dots\right)\mathbf{x}_0 \\ &= \mathbf{A}\left(\mathbf{I} + \mathbf{A}(t-t_0) + \frac{1}{2!}\mathbf{A}^2(t-t_0)^2 + \frac{1}{3!}\mathbf{A}^3(t-t_0)^3 \dots\right)\mathbf{x}_0 \\ &= \mathbf{A}e^{\mathbf{A}(t-t_0)}\mathbf{x}_0 \\ &= \mathbf{A}\mathbf{x}(t)\end{aligned}$$

## Example: Double Integrator

- ▶ Consider a second-order scalar LTI ODE:

$$\ddot{q} = u, \quad q(0) = q_0, \quad \dot{q}(0) = v_0$$

- ▶ It is called a **double integrator** because  $u(t)$  is integrated twice before it affects  $q$
- ▶ **State-space model:** let  $\mathbf{x} = (q, \dot{q})$ :

$$\dot{\mathbf{x}} = \underbrace{\begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix}}_{\mathbf{A}} \mathbf{x} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} \mathbf{u}, \quad \mathbf{x}(0) = \mathbf{x}_0 := \begin{bmatrix} q_0 \\ v_0 \end{bmatrix}$$

- ▶ Matrix exponential of  $\mathbf{A}$ :

$$\mathbf{A}^2 = \mathbf{0} \quad \Rightarrow \quad e^{\mathbf{A}t} = \mathbf{I} + \mathbf{A}t = \begin{bmatrix} 1 & t \\ 0 & 1 \end{bmatrix}$$

- ▶ When  $u \equiv 0$ , the solution of the double integrator system is:

$$\mathbf{x}(t) = e^{\mathbf{A}t} \mathbf{x}_0 = \begin{bmatrix} 1 & t \\ 0 & 1 \end{bmatrix} \begin{bmatrix} q_0 \\ v_0 \end{bmatrix} = \begin{bmatrix} q_0 + tv_0 \\ v_0 \end{bmatrix}$$

## Example: Undamped Oscillator

- ▶ Consider a spring-mass system with zero damping:

$$\ddot{q} + \omega_0^2 q = u, \quad q(0) = q_0, \quad \dot{q}(0) = v_0$$

- ▶ **State-space model:** let  $\mathbf{x} = (q, \dot{q}/\omega_0)$ :

$$\dot{\mathbf{x}} = \underbrace{\begin{bmatrix} 0 & \omega_0 \\ -\omega_0 & 0 \end{bmatrix}}_{\mathbf{A}} \mathbf{x} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} \mathbf{u}, \quad \mathbf{x}(0) = \mathbf{x}_0 := \begin{bmatrix} q_0 \\ v_0 \end{bmatrix}$$

- ▶ Matrix exponential of  $\mathbf{A}t$ :  $e^{\mathbf{A}t} = \begin{bmatrix} \cos(\omega_0 t) & \sin(\omega_0 t) \\ -\sin(\omega_0 t) & \cos(\omega_0 t) \end{bmatrix}$

- ▶ This can be verified by differentiation:

$$\frac{d}{dt} e^{\mathbf{A}t} = \begin{bmatrix} 0 & \omega_0 \\ -\omega_0 & 0 \end{bmatrix} \begin{bmatrix} \cos(\omega_0 t) & \sin(\omega_0 t) \\ -\sin(\omega_0 t) & \cos(\omega_0 t) \end{bmatrix} = \mathbf{A} e^{\mathbf{A}t}$$

- ▶ When  $u \equiv 0$ , the solution of the undamped oscillator is:

$$\mathbf{x}(t) = e^{\mathbf{A}t} \mathbf{x}_0 = \begin{bmatrix} \cos(\omega_0 t) & \sin(\omega_0 t) \\ -\sin(\omega_0 t) & \cos(\omega_0 t) \end{bmatrix} \begin{bmatrix} q_0 \\ v_0 \end{bmatrix}$$

## Where Does the Homogeneous LTI ODE Solution Come From?

- ▶ The solution to  $\dot{\mathbf{x}} = \mathbf{A}\mathbf{x}$  with  $\mathbf{x}(t_0) = \mathbf{x}_0$  should satisfy:

$$\mathbf{x}(t) = \mathbf{x}_0 + \int_{t_0}^t \mathbf{A}\mathbf{x}(\tau) d\tau$$

- ▶ This is an implicit equation. Replace the expression above into the integral:

$$\begin{aligned}\mathbf{x}(t) &= \mathbf{x}_0 + \int_{t_0}^t \mathbf{A} \left( \mathbf{x}_0 + \int_{t_0}^{\tau} \mathbf{A}\mathbf{x}(\tau_1) d\tau_1 \right) d\tau \\ &= (\mathbf{I} + \mathbf{A}(t - t_0)) \mathbf{x}_0 + \int_{t_0}^t \int_{t_0}^{\tau} \mathbf{A}^2 \mathbf{x}(\tau_1) d\tau_1 d\tau\end{aligned}$$

- ▶ Repeat the step above:

$$\begin{aligned}\mathbf{x}(t) &= (\mathbf{I} + \mathbf{A}(t - t_0)) \mathbf{x}_0 + \int_{t_0}^t \int_{t_0}^{\tau} \mathbf{A}^2 \left( \mathbf{x}_0 + \int_{t_0}^{\tau_1} \mathbf{A}\mathbf{x}(\tau_2) d\tau_2 \right) d\tau_1 d\tau \\ &= \left( \mathbf{I} + \mathbf{A}(t - t_0) + \frac{1}{2} \mathbf{A}^2 (t - t_0)^2 \right) \mathbf{x}_0 + \int_{t_0}^t \int_{t_0}^{\tau} \int_{t_0}^{\tau_1} \mathbf{A}^3 \mathbf{x}(\tau_2) d\tau_2 d\tau_1 d\tau \\ &= \dots = \left( \mathbf{I} + \mathbf{A}(t - t_0) + \frac{1}{2!} \mathbf{A}^2 (t - t_0)^2 + \frac{1}{3!} \mathbf{A}^3 (t - t_0)^3 + \dots \right) \mathbf{x}_0\end{aligned}$$



# LTI ODE Solution

## Theorem

*The linear time-invariant ordinary differential equation:*

$$\dot{\mathbf{x}} = \mathbf{A}\mathbf{x} + \mathbf{B}\mathbf{u}, \quad \mathbf{x}(t_0) = \mathbf{x}_0$$

*has a unique solution:*

$$\mathbf{x}(t) = e^{\mathbf{A}(t-t_0)}\mathbf{x}_0 + \int_{t_0}^t e^{\mathbf{A}(t-\tau)}\mathbf{B}\mathbf{u}(\tau)d\tau$$

## Proof

► Initial condition:

$$\mathbf{x}(t_0) = e^{\mathbf{A}(t_0-t_0)}\mathbf{x}_0 + \int_{t_0}^{t_0} e^{\mathbf{A}(t_0-\tau)}\mathbf{B}\mathbf{u}(\tau)d\tau = \mathbf{I}\mathbf{x}_0 + \mathbf{0} = \mathbf{x}_0$$

► ODE:

$$\begin{aligned}\frac{d}{dt}\mathbf{x}(t) &= \frac{d}{dt}\left(e^{\mathbf{A}(t-t_0)}\mathbf{x}_0\right) + \frac{d}{dt}\left(\int_{t_0}^t e^{\mathbf{A}(t-\tau)}\mathbf{B}\mathbf{u}(\tau)d\tau\right) \\ &= \mathbf{A}e^{\mathbf{A}(t-t_0)}\mathbf{x}_0 + \frac{d}{dt}\left(e^{\mathbf{A}t}\int_{t_0}^t e^{-\mathbf{A}\tau}\mathbf{B}\mathbf{u}(\tau)d\tau\right) \\ &= \mathbf{A}e^{\mathbf{A}(t-t_0)}\mathbf{x}_0 + (\mathbf{A}e^{\mathbf{A}t})\left(\int_{t_0}^t e^{-\mathbf{A}\tau}\mathbf{B}\mathbf{u}(\tau)d\tau\right) + (e^{\mathbf{A}t})(e^{-\mathbf{A}t}\mathbf{B}\mathbf{u}(t)) \\ &= \mathbf{A}\left(e^{\mathbf{A}(t-t_0)}\mathbf{x}_0 + \int_{t_0}^t e^{\mathbf{A}(t-\tau)}\mathbf{B}\mathbf{u}(\tau)d\tau\right) + \mathbf{B}\mathbf{u}(t) \\ &= \mathbf{A}\mathbf{x}(t) + \mathbf{B}\mathbf{u}(t)\end{aligned}$$

## LTI ODE Solution

- ▶ Consider the LTI ODE system:

$$\begin{aligned}\dot{\mathbf{x}} &= \mathbf{A}\mathbf{x} + \mathbf{B}\mathbf{u}, & \mathbf{x}(t_0) &= \mathbf{x}_0 \\ \mathbf{y} &= \mathbf{C}\mathbf{x} + \mathbf{D}\mathbf{u}\end{aligned}$$

- ▶ The system output satisfies the **convolution equation**:

$$\mathbf{y}(t) = \mathbf{C}e^{\mathbf{A}(t-t_0)}\mathbf{x}_0 + \int_{t_0}^t \mathbf{C}e^{\mathbf{A}(t-\tau)}\mathbf{B}\mathbf{u}(\tau)d\tau + \mathbf{D}\mathbf{u}(t)$$

- ▶ **Observations:**

- ▶ Due to the linearity of matrix-vector multiplication and integration, the output is **jointly linear** in the initial condition  $\mathbf{x}_0$  and the input  $\mathbf{u}(t)$
- ▶ The objective of control design is to choose an input signal  $\mathbf{u}(t)$  to shape the output  $\mathbf{y}(t)$ , e.g., to achieve regulation or tracking without overshoot or oscillations and with robustness to noise
- ▶ Using the convolution equation directly for control design can be challenging
- ▶ We will look for a simpler relationship between  $\mathbf{u}(t)$  and  $\mathbf{y}(t)$  by transforming the LTI ODE from the time domain to the frequency domain using a **Laplace transform**

# LTI Difference Equation Solution

## Theorem

The linear time-invariant difference equation:

$$\mathbf{x}_{k+1} = \mathbf{A}\mathbf{x}_k + \mathbf{B}\mathbf{u}_k$$

has a unique solution:

$$\mathbf{x}_k = \mathbf{A}^k \mathbf{x}_0 + \sum_{j=0}^{k-1} \mathbf{A}^{k-j-1} \mathbf{B}\mathbf{u}_j$$

**Proof:**

- ▶ **Base case** (time  $k = 1$ ):  $\mathbf{x}_1 = \mathbf{A}\mathbf{x}_0 + \mathbf{B}\mathbf{u}_0$
- ▶ **Induction hypothesis** (time  $k$ ):  $\mathbf{x}_k = \mathbf{A}^k \mathbf{x}_0 + \sum_{j=0}^{k-1} \mathbf{A}^{k-j-1} \mathbf{B}\mathbf{u}_j$
- ▶ **Induction step** (time  $k + 1$ ):

$$\begin{aligned} \mathbf{x}_{k+1} &= \mathbf{A}\mathbf{x}_k + \mathbf{B}\mathbf{u}_k = \mathbf{A} \left( \mathbf{A}^k \mathbf{x}_0 + \sum_{j=0}^{k-1} \mathbf{A}^{k-j-1} \mathbf{B}\mathbf{u}_j \right) + \mathbf{B}\mathbf{u}_k \\ &= \mathbf{A}^{k+1} \mathbf{x}_0 + \sum_{j=0}^{k-1} \mathbf{A}^{k-j} \mathbf{B}\mathbf{u}_j + \mathbf{B}\mathbf{u}_k = \mathbf{A}^{k+1} \mathbf{x}_0 + \sum_{j=0}^k \mathbf{A}^{k-j} \mathbf{B}\mathbf{u}_j \end{aligned}$$